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Advanced Signal Processing Techniques for Fault Detection and Diagnosis in a Wind Turbine Induction Generator Drive Train: A Comparative Study

E. Al Ahmar, V. Choqueuse, M.E.H. Benbouzid, Y. Amirat, J. El Assad, R. Karam, and S. Farah

Abstract—This paper deals with the diagnosis of Wind Turbines based on generator current analysis. It provides a comparative study between traditional signal processing methods, such as periodograms, with more sophisticated approaches. Performances of these techniques are assessed through simulation experiments and compared for several types of fault, including air-gap eccentricities, broken rotor bars and bearing damages.

Index Terms—Wind turbines, motor current signature analysis, time-frequency signal processing methods, wavelet analysis, failure diagnosis.

I. INTRODUCTION

Due to environmental considerations and to the exhaustion of fossil resources, renewable energy sources are gaining more and more importance these days. Since 2004, many indicators of the development of renewable energy have been constantly rising. For example, since 2004, annual renewable energy investment has increased fourfold to reach \$120 billion in 2008 [1]. Wind power has a major share in this increase since, from 2004 to 2008, wind power capacity increased by 250% to attend a value of 121GW (see Fig.1 [1]). Due to different factors, such as wind speed and acoustic noise, wind parks are being mainly constructed offshore. Studies shows that offshore wind power produced 1873MW in Europe. To increase this production, 11 parks are currently under construction and 26 parks are in project (see table I). Due to the difficulties of maintenance in offshore environment, teledetection of wind turbine faults is becoming a crucial issue. Different methods of fault detection exist. These include monitoring of the acoustic vibration, internal and external temperature, stator voltages and/or currents. Although thermal and vibration monitoring have been utilized for decades, most of the recent research has been directed toward electrical monitoring, with emphasis on the generator stator current. In this study, the generator stator

Fig. 1. Annual wind power in the world between years 1996 and 2008.

TABLE I
WIND POWER PRODUCED OFFSHORE IN EUROPE [1] [4].

	Operational	Construction	Project
Number of parks	28	11	26
Produced Power (MW)	1873	2743	5716

current is analyzed through signal processing tools for fault detection [2], [3]. The performances of several signal processing tools for the detection of air-gap eccentricity, broken rotor bars and bearing damages are compared. This paper is organized as follows. Section II presents the considered faults, section III describes the employed signal processing techniques and section IV provides an experimental comparison of these methods.

II. MAJOR FAULTS STUDIED

Wind turbine is a complex mechanical system. Its major components are presented in Fig. 2. Faults can occur almost anywhere in this system; examples of mechanical faults range from a bent shaft, which results in a rub between the rotor and stator, to a shorted rotor winding or gearbox failures. This paper focuses on faults that leads to air-gap eccentricities, broken rotor bars and bearing damages.

A. Air gap eccentricity

Eccentricity faults in electrical machines are caused by an unequal air gap between the stator and rotor [5]–[7]. When

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Fig. 2. Wind turbine nacelle cross-section.

Fig. 3. Mechanical representation of bearings.

this air gap increases the unbalanced magnetic pull or radial forces can cause the stator and rotor rubbing, which may result in major damage in both the stator and the rotor. Two types of eccentricity exist or even tend to coexist with each others. The first is the static eccentricity where the minimal radial air gap length is fixed in space. It may be caused by an oval shape of the stator core or by a misalignment of the stator and rotor. Dynamic eccentricity, on the other hand, is caused by a difference between the center of the rotor and the center of rotation which may be caused by a bent rotor shaft or a bearing fault, etc. both these eccentricities cause a rub between the stator and the rotor causing critical damage in both of them which may lead to a total breakdown of the machine. In order to detect these eccentricity faults, two methods exist. The first one consists of monitoring the behavior of the current at the sidebands of the slot frequencies. The main drawback of this approach relies on the fact that it requires knowledge of the intimate details of the machine's anatomy [8]. The second method is based on the monitoring of the current at the fundamental sidebands of the supply frequency. Frequencies of interest are located at f_{ecc} given by the following equation [8]:

$$f_{ecc} = f_s \left[1 \pm m \left(\frac{1-s}{p} \right) \right] \quad (1)$$

where f_s is the electrical supply frequency, s is the per-unit slip, p is the number of poles and $m \in \mathbb{N}$.

B. Broken rotor Bars

Though several types of rotor cages exist, most wind turbines have cast rotors. In this case, damage of a rotor bar is fatal since it cannot be repaired. Broken rotor bars can be caused by thermal, magnetic, environmental or mechanical stresses [9]. When a rotor bar breaks in the rotor cage, it induces magnetic anomalies which produce harmonic frequencies in the current spectrum at frequencies f_{brb} [9]:

$$f_{brb} = f_s \left[k \left(\frac{1-s}{p} \right) \pm s \right] \quad (2)$$

where $k/p = 1, 3, 5, 7, 11, 13$.

C. Bearing Damage

Most of rotating mechanical systems contains one or many bearings to insure smooth rotation with minimal losses. Bearings consist of an inner ring and external one, between which rotates a number of balls (see Fig.3). Bearing may be damaged by sudden load variations or rotor-stator misalignments. Bearing faults are the most common fault in wind turbines (40-50%). Vibrations associated with bearing faults reflect themselves in the current spectrum at frequencies [9]:

$$f_{bng} = |f_s \pm m f_v| \quad (3)$$

where $m \in \mathbb{N}$ and f_v depends on bearing fault. Bearing faults include

- Faults caused by an outer bearing race defect where:

$$f_v = \frac{\alpha f_r}{2} \left[1 - \frac{b_d}{p_d} \cos(\beta) \right] \quad (4)$$

- Faults caused by an inner bearing race defect where:

$$f_v = \frac{\alpha f_r}{2} \left[1 + \frac{b_d}{p_d} \cos(\beta) \right] \quad (5)$$

- Faults caused by a ball defect where:

$$f_v = \frac{p_d f_r}{2 b_d} \left[1 + \left(\frac{b_d}{p_d} \cos(\beta) \right)^2 \right] \quad (6)$$

- Faults caused by a train defect where frequencies of interest are given by:

$$f_v = \frac{f_r}{2} \left[1 - \frac{b_d}{p_d} \cos(\beta) \right] \quad (7)$$

where f_r is the rotational frequency, α is the number of balls, b_d and p_d are respectively the ball diameter and the bearing pitch diameter (see Fig.2) and β is the contact angle of the ball.

III. ADVANCED SIGNAL PROCESSING TECHNIQUES

This section presents different signal processing technique which can be employed to extract the frequency content of a discrete signal. In this following, the current will be denoted by the discrete signal $x[n]$, which is obtained by sampling the continuous time current every $T_s = 1/F_s$ seconds.

A. Periodogram

The periodogram, $P_x(f)$, estimates the Power Spectral Density (PSD) of a signal $x[n]$. It is given by [10]:

$$P_x(f) = \frac{|X(f)|^2}{N} \quad (8)$$

where $X(f)$ is the Discrete Fourier Transform of $x[n]$ i.e.

$$X(f) = \sum_{n=0}^{N-1} x[n] e^{-2j\pi f n / F_s} \quad (9)$$

where F_s is the sampling frequency. In practice, (9) is evaluated through a Fast Fourier Transform (FFT) [10], which reduces the number of computations to $\mathcal{O}(N \log(N))$ operations. One should note The periodogram is not a consistent estimator of the PSD since it has a non-zero bias and its variance does not tend to zero as the data length N tends to infinity. Despite of this drawback, the periodogram has been used extensively for failure detection in the literature [8].

B. Welch Periodogram

As compared to the classical periodogram, Welch periodogram is an improved estimator of the PSD that reduces both the variance and the bias. The Welch method divides $x[n]$ into segments, computes a modified periodogram of each segment and then averages the result. The Welch periodogram, $P_w(f)$, can be expressed into a mathematical form as [10]:

$$P_w(f) = \frac{1}{L} \sum_{k=1}^{k=L} P_{xw}^{(k)}(f) \quad (10)$$

where:

$$P_{xw}^{(k)}(f) = \frac{|X_{xw}^{(k)}(f)|^2}{NU} \quad (11)$$

and where U is a normalization factor. Furthermore, $X_w^{(k)}(f)$ corresponds to the DFT of the windowed signal $x[n]w[n-\tau_k]$, where $w[\cdot]$ is a time-window (hanning, hamming, kaiser) and where τ_k is a time lag.

C. Spectrogram

To obtain the evolution of the frequency content over time, Fourier transforms can be computed for different time segments and then arranged one next to the other over the time axis. This method is known as the Short Time Fourier Transform (STFT). For discrete signals, the STFT is given by [11]

$$STFT[f, \tau] = \sum_{n=0}^{N-1} x[n] w[n-\tau] e^{-2j\pi f n / F_s} \quad (12)$$

where $w[\cdot]$ is a time window. The spectrogram is defined as the square modulus of the STFT i.e. $|STFT[f, \tau]|^2$. The time and frequency resolution is limited by the Heisenberg-Gabor inequality [11]. In the case of the spectrogram, this resolution is the same for all time-frequency bins. One should note that the spectrogram have been used previously in [12], [13] for diagnosis purpose in time-varying condition.

Fig. 4. Simulation of the current of a healthy machine in blue and of a machine with an eccentricity problem between 1.024s and 2.024s.

D. Scalogram

Wavelet Transform (WT) provides a time-scale (or time-frequency) representation of a signal. While STFT gives constant time-frequency resolution, WT is a multi-resolution technique which analyses frequencies with different resolutions. The WT gives a good time resolution and poor frequency resolution at high frequencies, and provides good frequency resolution and poor time resolution at low frequencies [14]. The Wavelet Transform at frequency scale l is given by:

$$W(a^l, \tau) = \sum_{n=0}^{N-1} x[n] \psi_l^*[n - \tau] \quad (13)$$

where

$$\psi_l[n] = \frac{1}{\sqrt{a^l}} \psi\left(\frac{n}{a^l}\right) \quad (14)$$

and where $\psi(\cdot)$ is the mother wavelet which satisfies a number of conditions. The scalogram is defined as the square modulus of the WT i.e. $|W(a^l, \tau)|^2$. One should note that the Wavelet Transform have been used previously in [15]–[17] for failure detection.

IV. SIMULATION RESULTS

The focus of this section is to find which signal processing method is the most suitable for the detection of each mechanical fault. In this section, the simulated signal $x[n]$ is given by:

$$x[n] = \sin(2\pi f_s n / F_s) + e[n] + b[n] \quad (15)$$

where $F_s = 1000$ Hz, $e[n]$ depends on the fault and $b[n]$ is a gaussian noise with zero mean and a variance equals to $\sigma^2 = 10^{-4}$ i.e. $b[n] \sim \mathcal{N}(0, 10^{-4})$. For all simulations, an hamming window $w[n]$ have been used for the Welch periodogram and the spectrogram. For the Welch periodogram, eight sections of equal length, each with 50% overlap, have been extracted before averaging. Finally, the length of the spectrogram window, $w[\cdot]$, have been fixed to 256 samples and the mother wavelet of the scalogram was a Mexican hat.

Fig. 5. Periodogram of the monitored current in the case of an eccentricity fault between 1.024s and 2.024s.

Fig. 6. Welch periodogram of the monitored current in the case of an eccentricity fault between 1.024s and 2.024s.

A. Air-gap eccentricity

In this subsection, air gap eccentricity faults are simulated by generating an additive signal $e[n]$, whose spectral components are given by (1). The parameters are fixed to $s = 0.033$, $p = 2$, $m = [1, \dots, 5]$ and the eccentricity fault is simulated between 1.024s and 2.024s (see Fig. 4). The first signal processing method applied to these signals is the periodogram (see Fig. 5). The five harmonics ($m = [1, \dots, 5]$), corresponding to the fault are theoretically equal to 89.01Hz, 118.02Hz, 147.03Hz, 176.04Hz and 205.5Hz. Using the classical periodogram technique, these frequencies are estimated at 89.43Hz, 118.7Hz, 148Hz, 177.3Hz and 206.6Hz respectively. The same results are obtained with the Welch periodogram (see Fig. 6). Figures 7 and 8 display the spectrogram and the scalogram respectively. One can observe that these methods bring new information that could not be extracted with the previous periodograms. Furthermore, one can note that the scalogram has advantages over the spectrogram since it provides a multiresolution analysis of the signal.

B. Broken rotor bars

The second fault studied is a broken rotor bar fault. Its signature is given by equation (2). Figures 9 and 10 show the classical and Welch periodogram of the current in the case of

Fig. 7. Spectrogram of the monitored current in the case of an eccentricity fault between 1.024s and 2.024s.

Fig. 8. Scalogram of the monitored current in the case of an eccentricity fault between 1.024s and 2.024s.

TABLE II
HARMONICS DETECTED IN THE CASE OF A BROKEN ROTOR BAR.

s	harmonics $f_s(1 - 2s)$			
	1	2	3	4
f_{brb}	55.98 Hz	171.9 Hz	287.9 Hz	404.2 Hz
Per./Welch	56.04 Hz	172.08 Hz	288.12 Hz	404.16 Hz
Error	0.107%	0.104%	0.076%	0.01%

s	harmonics $f_s(1 + 2s)$			
	1	2	3	4
f_{brb}	60 Hz	176.04 Hz	292.08 Hz	408.12 Hz
Per./Welch	59.98 Hz	176.3 Hz	292.2 Hz	408.2 Hz
Error	0.033%	0.147%	0.041%	0.019%

a damaged rotor. Table II shows the frequency location of the harmonics associated with the broken bar. The harmonics are detected at the same frequencies for the classical and Welch periodogram. However, as one can observe from Fig. 10, the Welch periodogram presents a better signal to noise ratio. Figures 11 and 12 show the spectrogram and the scalogram, respectively. Despite the fact that this method brings no real amelioration in the spectral resolution, it brings great information on the time occurrence of the fault, which can be exploited for fault discrimination.

Fig. 9. Periodogram of the monitored current in the case of broken rotor bars between 1.024s and 2.024s.

Fig. 12. Scalogram of the monitored current in the case of broken rotor bars between 1.024s and 2.024s.

Fig. 10. Welch periodogram of the monitored current in the case of broken rotor bars between 1.024s and 2.024s.

Fig. 13. Periodogram of the monitored current in the case of bearing damages between 1.024s and 2.024s.

Fig. 11. Spectrogram of the monitored current in the case of broken rotor bars between 1.024s and 2.024s.

Fig. 14. Welch periodogram of the monitored current in the case of bearing damages between 1.024s and 2.024s.

C. Bearing damage

In wind turbine, bearings faults are the most common fault. In this case, the damage can be caused by different sources such as the outer or the inner rings, the balls or the race. In the following, all these causes are combined together to simulate the worst scenario. The four bearing fault signatures in (4), (5), (6) and (7) are included in the faulty signal. The other simulation parameters are fixed as follows: $m = \{1, 2\}$, $s=0.033$, $fr=29.01\text{Hz}$, $bd=0.01\text{m}$, $pd=0.1\text{m}$, $N=12$

and $\beta = 25^\circ$. The periodogram and the Welch periodogram are shown in Fig.13 and 14, respectively. As previously observed, the Welch periodogram leads to a better frequency representation than the classical periodogram in terms of signal to noise ratio. Figures 15 and 16 display the spectrogram and the scalogram of the signal. As previously discussed, the scalogram provides a more natural representation since low-frequency components have high-frequency resolution and high-frequency components have high-time resolution.

Fig. 15. Spectrogram of the monitored current in the case of bearing damages between 1.024s and 2.024s.

Fig. 16. Scalogram of the monitored current in the case of bearing damages between 1.024s and 2.024s.

V. CONCLUSION

This work presented a comparison between four signal processing methods for fault detection in the case of wind turbines. Experimental simulations have shown that the classical and Welch Periodogram have very close performances. However for noisy data, the Welch periodogram exhibits a better signal to noise ratio. As compared to Periodogram and Welch Periodogram, the spectrogram and the scalogram, which are time-frequency representations, bring up major information concerning the time occurrence of the fault. Future works

will focus on the evaluation of the presented techniques with experimental signals.

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